



10MHz, Precision, Rail-to-Rail I/O CMOS Operational Amplifier

FEATURES

- HIGH GAIN BANDWIDTH: 10MHz
- RAIL-TO-RAIL INPUT AND OUTPUT ±0.5mV Max Vos (RS8602)
- INPUT VOLTAGE RANGE: -0.2V to +5.7V with Vs = 5.5V
- SUPPLY RANGE: +2.5V to +5.5V
- SPECIFIED UP TO +125°C
- Micro SIZE PACKAGES: SOIC8, MSOP8

APPLICATIONS

- SENSORS
- ACTIVE FILTERS
- TEST EQUIPMENT
- DRIVING A/D CONVERTERS
- PHOTODIODE AMPLIFICATION

DESCRIPTION

The RES8602IDR,RES8602IMR families of products offer low voltage operation and rail-to-rail input and output, as well as excellent speed/power consumption ratio, providing an excellent bandwidth (10MHz) and slew rate of 6V/us. The op-amps are unity gain stable and feature an ultra-low input bias current.

The RES8602IDR,RES8602IMR has lower offset, which is guaranteed not upper than 0.5 mV at $25 ^{\circ}\text{C}$ with Vs = 5 V, $V_{\text{CM}} = V \text{s}/2$.

The devices are ideal for sensor interfaces, active filters and portable applications. The RES8602IDR RES8602IMR families of operational amplifiers are specified at the full temperature range of -40°C to +125°C under single or dual power supplies of 2.5V to 5.5V.

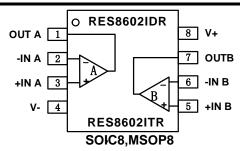
Device Information (1)

PART NUMBER	PACKAGE	BODY SIZE(NOM)
RES8602IDR	SOIC8	4.90mm×3.90mm
RES8602IMR	MSOP8	3.00mm×3.00mm

⁽¹⁾ For all available packages, see the orderable addendum at the end of the data sheet.



Pin Configuration and Functions (Top View)



Pin Description

NAME	PIN SOIC8,MSOP8	I/O	DESCRIPTION
-INA	2		Inverting input, channel A
+INA	3	1	Noninverting input, channel A
-INB	6	- 1	Inverting input, channel B
+INB	5	- 1	Noninverting input, channel B
OUTA	1	0	Output, channel A
OUTB	7	0	Output, channel B
V-	4	-	Negative (lowest) power supply
V+	8	-	Positive (highest) power supply

SPECIFICATIONS

Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) (1)

		MIN	MAX	UNIT
	Supply, Vs=(V+) - (V-)		7	
Voltage	Signal input pin (2)	(V-)-0.5	(V+) + 0.5	V
	Signal output pin (3)	(V-)-0.5	(V+) + 0.5	
	Signal input pin (2)	-10	10	mA
Current	Signal output pin (3)	-150	150	mA
	Output short-circuit (4)	Contir	nuous	
	Operating range, T _A	-40	125	
Temperature	Junction, T _J	-40	150	°C
	Storage, T _{stg}	-65	150	

⁽¹⁾ Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

ESD Ratings

			VALUE	UNIT
V _(ESD) Electrostatic discharge	Human-body model (HBM)	±3000	\/	
	Electrostatic discharge	Machine Model (MM)	±200	V

Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage Vs= (V+) - (V-)	Signal-supply	2.5		5.5	V
Supply voltage, Vs= (V+) - (V-)	Dual-supply	±1.25		±2.75	V

⁽²⁾ Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

⁽³⁾ Output terminals are diode-clamped to the power-supply rails. Output signals that can swing more than 0.5V beyond the supply rails should be current-limited to ±150mA or less.

⁽⁴⁾ Short-circuit to ground, one amplifier per package.



Thermal Information: RES8602IDR,RES8602IMR

THERMAL METRIC		RS8602IDR,RS8602IMR				
			UNIT			
		SOIC8	MSOP8			
R _{OJA}	Junction-to-ambient thermal resistance	116	165		°C/W	
R _{OJC(top)}	Junction-to-case(top) thermal resistance	60	53		°C/W	
R _{OJB}	Junction-to-board thermal resistance	56	87		°C/W	
Ψ _{JT}	Junction-to-top characterization parameter	12.8	4.9		°C/W	
Ψ_{JB}	Junction-to-board characterization parameter	98.3	85		°C/W	
R _{OJC(bot)}	Junction-to-case(bottom) thermal resistance	N/A	N/A		°C/W	

PACKAGE/ORDERING INFORMATION

Orderable Device	Package Type	Pin	Channel	Op Temp(°C)	Device Marking	Package Qty
RES8602IDR	SOIC8	8	2	-40°C ~125°C	RES8602IDR	Tape and Reel,4000
RES8602IMR	MSOP8	8	2	-40°C ~125°C	RES8602IMR	Tape and Reel,4000

NOTE:

(1) Equivalent to SOIC8,MSOP8.



ELECTRICAL CHARACTERISTICS

(At T_A = +25°C, Vs=5V, R_L = 10k Ω connected to V_S/2, and V_{OUT} = V_S/2, unless otherwise noted.)

		00115151010	_	RE	ES8602I	DR,RES	ES8602IMR	
	PARAMETER	CONDITIONS	ΤJ	MIN	TYP	MAX	UNIT	
POWER	SUPPLY							
Vs	Operating Voltage Range		25°C	2.5		5.5	V	
IQ	Quiescent Current/Amplifier		25°C		1.1	1.55	mA	
PSRR	Dower Cupply Dejection Datio	Vs=2.5V to 5.5V,	25°C	75	97		dB	
FORK	Power-Supply Rejection Ratio	V _{CM} =(V-)+0.5V	-40°C to 125°C	65			uБ	
ton	Turn-on Time		25°C		12		us	
INPUT								
		V _{CM} = V _S /2, RS721P	25°C	-0.5	±0.2	0.5	mV	
Vos	Input Offset Voltage	V _{CM} = V _S /2, RS722P	25°C	-0.5	±0.2	0.5	mV	
		V _{CM} = V _S /2, RS724P	25°C	-0.8	±0.3	0.8	mV	
Vos T _C	Input Offset Voltage Average Drift		-40°C to125°C		±2.6		uV/°C	
IB	Input Bias Current		25°C		±1	±10	pА	
los	Input Offset Current		25°C		±1	±10	pА	
Vсм	Common-Mode Voltage Range	Vs= 5.5V	25°C	-0.2		5.7	V	
		Vs= 5.5V,	25°C	77	97			
CMRR	Common-Mode Rejection Ratio	V _{CM} =-0.2V to 4V	-40°C to 125°C	70			dB	
CIVIKK	Common-wode Rejection Ratio	Vs= 5.5V,	25°C	65	82		uБ	
		V _{CM} =-0.2V to 5.7V	-40°C to 125°C	60				
OUTPUT	7							
		R _L =2KΩ, Vo=	25°C	86	105			
Aol	Open-Loop Voltage Gain	0.15V to 4.85V	-40°C to 125°C	65			dB	
AOL	Open-Loop Voltage Gain	$R_L=10K\Omega$, Vo=	25°C	96	110		ub	
		0.05V to 4.95V	-40°C to 125°C	75				
	Output Swing From Rail	R _L =2KΩ	25°C		52		mV	
	Output Swing From Kaii	R _L =10KΩ	25 C		7		IIIV	
I _{OUT}	Output Current Source		25°C		140		mA	
FREQUE	ENCY RESPONSE							
SR	Slew Rate		25°C		6		V/us	
GBP	Gain-Bandwidth Product		25°C		10		MHz	
PM	Phase Margin		25°C		62		0	
ts	Setting Time,0.1%		25°C		0.5		us	
	Overload Recovery Time	V _{IN} -Gain≥V _S	25°C		3.2		us	
NOISE								
ο.	Input Voltage Noise Density	f = 1KHz	25°C		9.5		nV/√Hz	
еn	input voitage Noise Delisity	f = 10KHz	25°C		6.5		nV/√Hz	



At $T_A = +25$ °C, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.

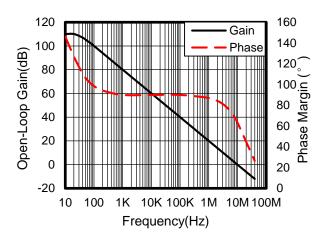


Figure 1. Open-loop Gain and Phase vs Frequency

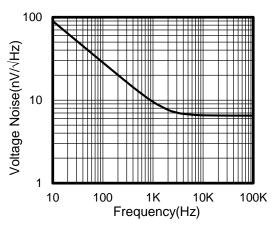


Figure 3. Input Voltage Noise Spectral Density vs Frequency

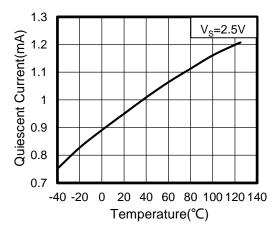


Figure 5. Quiescent Current vs Temperature

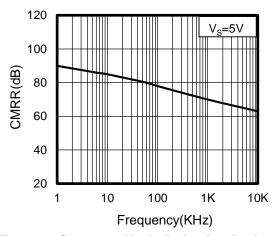


Figure 2. Common-Mode Rejection Ratio vs Frequency

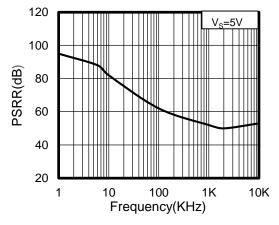


Figure 4. Power-Supply Rejection Ratio vs Frequency

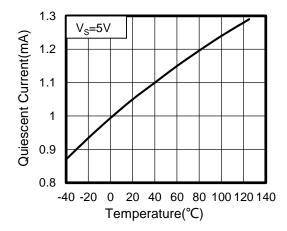


Figure 6. Quiescent Current vs Temperature



At $T_A = +25$ °C, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.

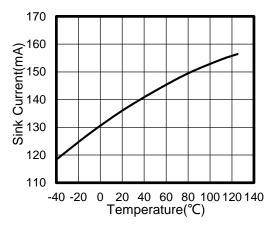


Figure 7. Sink Current vs Temperature

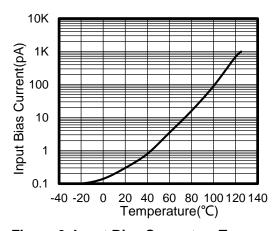


Figure 9. Input Bias Current vs Temperature

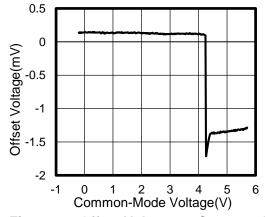


Figure 11. Offset Voltage vs Common-Mode Voltage

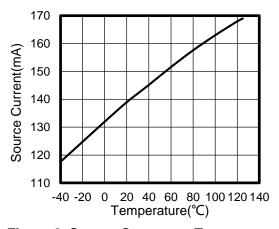


Figure 8. Source Current vs Temperature

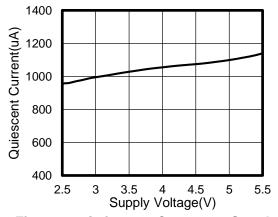
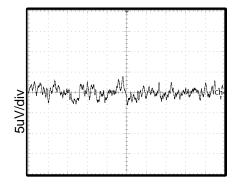


Figure 10. Quiescent Current vs Supply Voltage



Time(1s/div)

Figure 12. 0.1Hz to 10Hz Input Voltage Noise



At $T_A = +25^{\circ}C$, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.

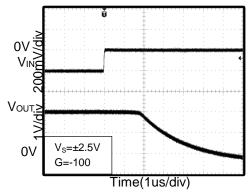


Figure 13. Positive Overvoltage Recovery

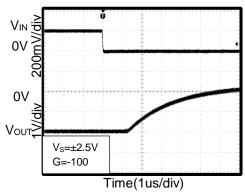


Figure 14. Negative Overvoltage Recovery

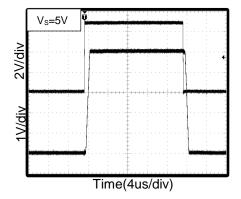


Figure 15. Large-Signal Step Response

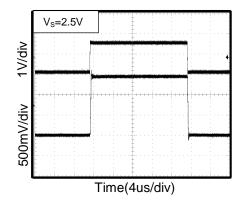


Figure 16. Large-Signal Step Response

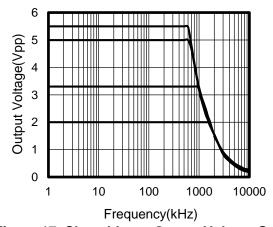


Figure 17. Closed-Loop Output Voltage Swing

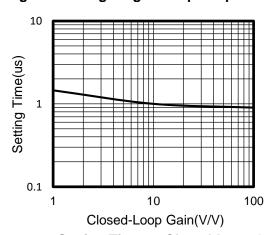


Figure 18. Setting Time vs Closed-Loop Gain



At $T_A = +25$ °C, $V_S=5V$, $R_L = 10k\Omega$ connected to $V_S/2$, $V_{OUT} = V_S/2$, unless otherwise noted.

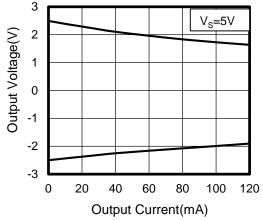


Figure 19. Output Voltage vs Output Current

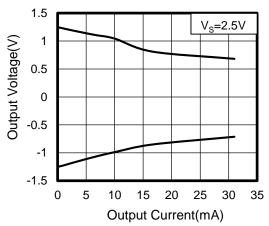


Figure 20. Output Voltage vs Output Current

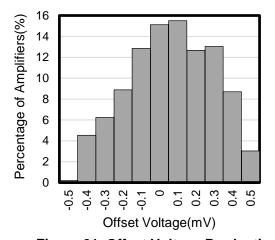


Figure 21. Offset Voltage Production Distribution



Detailed Description

Overview

The RES8602 devices are unity-gain stable, dual and qual-channel op amps with low noise and distortion. The device consists of a low noise input stage with a folded cascade and a rail-to-rail output stage. This topology exhibits superior noise and distortion performance across a wide range of supply voltages that are not delivered by legacy commodity audio operational amplifiers.

Phase Reversal Protection

The RES8602 family has in tern alphase e-reversal protection. Many op amps exhibit phase reversal when the input is driven beyond the linear common-mode range. This condition is most often encountered in noninverting circuits when the input is driven beyond the specified common-mode voltage range, causing the output to reverse into the opposite rail. The input of the RS8602 prevents phase revers a lwith excessive common-mode voltage. Instead, the appropriate rail limits the output voltage. This performance is shown in figure

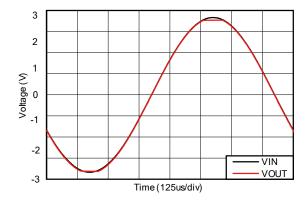


Figure 22. Output Waveform Devoid of Phase Reversal During an Input Overdrive Condition

EMI Rejection Ratio (EMIRR)

The electromagnetic interference (EMI) rejection ratio, or EMIRR, describes the EMI immunity of operational amplifiers. An adverse effect that is common to many operational amplifiers is a change in the offset voltage as a result of RF signal rectification. An operational amplifier that is more efficient at rejecting this change in offset as a result of EMI has a higher EMIRR and is quantified by a decibel value. Measuring EMIRR can be performed in many ways, but this document provides the EMIRR IN+, which specifically describes the EMIRR performance when the RF signal is applied to the noninverting input pin of the operational amplifier. In general, only the noninverting input is tested for EMIRR for the following three reasons:

- Operational amplifier input pins are known to be the most sensitive to EMI, and typically rectify RF signals better than the supply or output pins.
- The noninverting and inverting operational amplifier inputs have symmetrical physical layouts and exhibit nearly matching EMIRR performance.
- EMIRR is easier to measure on noninverting pins than on other pins because the noninverting input pin can be isolated on a printed-circuit-board (PCB). This isolation allows the RF signal to be applied directly to the noninverting input pin with no complex interactions from other components or connecting PCB traces.



Detailed Description (continued)

The EMIRR IN+ of the RES8602 is plotted vers us f requency in Figure. If available, any dual and quad operational amplifier device versions have approximately identical EMIRR IN+ performance. The RES8602 unity gain bandwidth is 10MHz. EMIRR performance below this frequency denotes interfering signals that fall within the operational amplifier bandwidth.

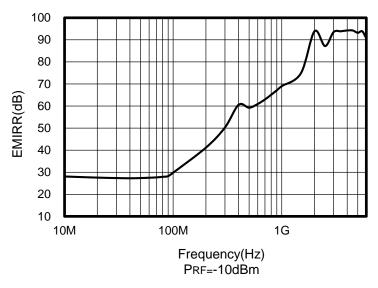


Figure 23. RES8602 EMIRR vs Frequency

EMIRR IN+ Test Configuration

Figure 24 shows the circuit configuration for testing the EMIRR IN+. An RF source is connected to the operational amplifier noninverting input pin using a transmission line. The operational amplifier is configured in a unity-gain buffer topology with the output connected to a low-pass filter (LPF) and a digital multimeter (DMM). A large impedance mismatch at the operational amplifier input causes a voltage reflection; however, this effect is characterized and accounted for when determining the EMIRR IN+. The resulting dc offset voltage is sampled and measured by the multimeter. The LPF isolates the multimeter from residual RF signals that can interfere with multimeter accuracy.

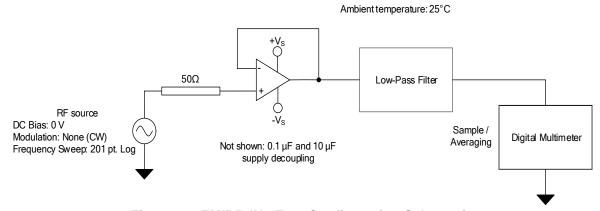


Figure 24. EMIRR IN+ Test Configuration Schematic



APPLICATION NOTE

The RES8602 are high precision, rail -to-rail operational amplifiers that can be run from a single-supply voltage 2.5V to 5.5V (± 1.25 V to ± 2.75 V). Supply voltages higher than 7V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Good layout practice mandates use of a 0.1uF capacitor place closely across the supply pins.

Typical Applications 25-kHz Low-pass Filter

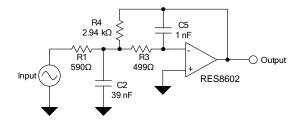


Figure 25. 25-kHz Low-Pass Filter

Design Requirements

Low-pass filters are commonly employed in signal processing applications to reduce noise and prevent aliasing. The RES8602 devices are ideally suited to construct high-speed, high-precision active filters. Figure 25 shows a second-order, low-pass filter commonly encountered in signal processing applications. Use the following parameters for this design example:

- Gain = 5 V/V (inverting gain)
- Low-pass cutoff frequency = 25 kHz
- Second-order Chebyshev filter response with 3-dB gain peaking in the passband

Detailed Design Procedure

The infinite-gain multiple-feedback circuit for a low-pass network function is shown in Figure 25. Use Equation 1 to calculate the voltage transfer function.

$$\frac{\text{Output}}{\text{Input}}(s) = \frac{-1/R_1R_3C_2C_5}{s^2 + (s/C_2)(1/R_1 + 1/R_3 + 1/R_4) + 1/R_3R_4C_2C_5}$$
(1)

This circuit produces a signal inversion. For this circuit, the gain at dc and the low-pass cutoff frequency are calculated by Equation 2:

Gain =
$$\frac{R_4}{R_1}$$

 $f_C = \frac{1}{2\pi} \sqrt{(1/R_3 R_4 C_2 C_5)}$

Application Curve

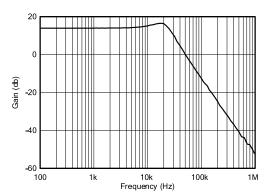


Figure 26. Low-pass filter transfer function



LAYOUT

Layout Guideline

Attention to good layout practices is always recommended. Keep traces short. When possible, use a PCB ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1uF capacitor closely across the supply pins.

These guidelines should be applied throughout the analog circuit to improve performance and provide benefits such as reducing the EMI susceptibility.

Layout Example

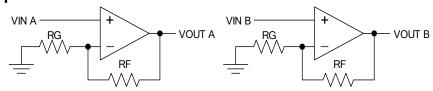


Figure 27. Schematic Representation

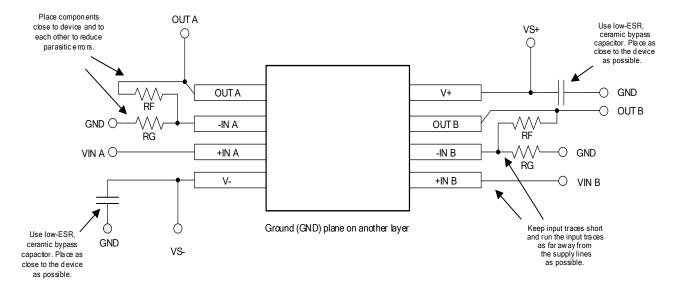
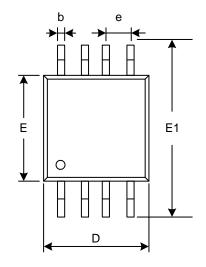
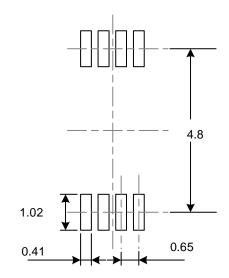


Figure 28. Layout Example

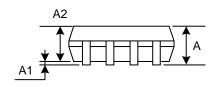


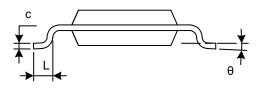
MSOP8





RECOMMENDED LAND PATTERN (Unit: mm)

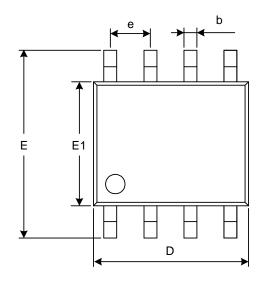


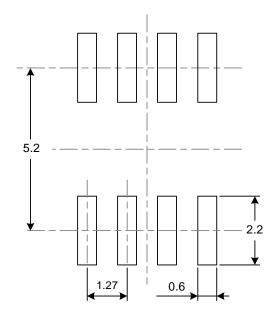


Cymphol	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Min Max		Max	
А	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.006	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
е	0.650	(BSC)	0.026(BSC)		
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
L	0.400	0.800	0.016	0.031	
θ	0°	6°	0°	6°	

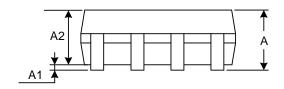


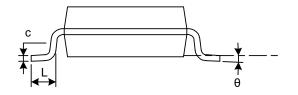
SOIC8





RECOMMENDED LAND PATTERN (Unit: mm)



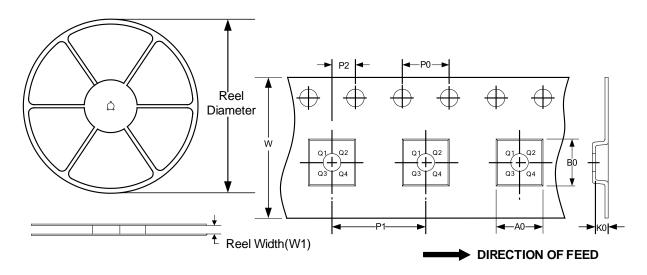


Cymhal	Dimensions	In Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
А	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.007	0.010	
D	4.800	5.000	0.189	0.197	
е	1.270	(BSC)	0.050(BSC)		
E	5.800	6.200	0.228	0.244	
E1	3.800	4.000	0.150	0.157	
L	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	



TAPE AND REEL INFORMATION REEL DIMENSIONS

TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width(mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOIC8	13"	12.4	6.40	5.40	2.10	4.0	8.0	2.0	12.0	Q1
MSOP8	13"	12.4	5.20	3.30	1.50	4.0	8.0	2.0	12.0	Q1